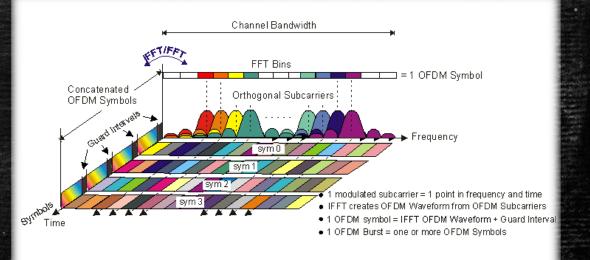
## Introduction to OFDM

**Basic concepts** 



## Definitions

- Orthogonal Frequency Division Multiplexion
  - A number of orthogonal carriers bear the information
  - Resilient to time synchronization errors
  - Very high spectral efficiency
- OFDM is a wide used transmission technique
   ADSL
  - IEEE 802.11a/g/n and WiMAX
  - DVB-T, DVB-H, DVB-T<sub>2</sub>, DVB-NGH.



The energy of a signal is defined as

$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

There are signals with E = ∞, as periodic signals for them it is defined the average power:

$$P = \lim_{T_0 \to \infty} \frac{1}{2T_0} \int_{-T_0}^{T_0} |x(t)|^2 dt$$

- Signals can be classified in
  - Energy signals:  $0 < E < \infty$
  - Power signals:  $0 < P < \infty$

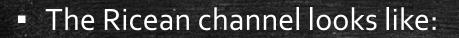
## Propagation scenarios

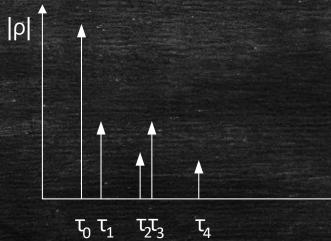
- In wireless communications there are much harder propagation conditions than in wired communications
- There are different propagations paths from the transmitter to the receiver (multi path propagation)

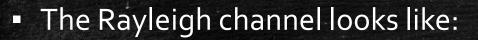


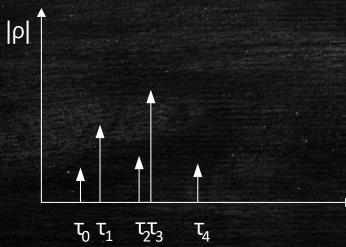
- The propagation scenarios are usually modelled as a FIR filter which taps represent:
  - Delay (τ)
  - Path complex gain (ρ)

- The propagation scenarios can be divided in two main types
  - Ricean (there is line of sight and hence direct ray)
  - Rayleigh (there is no line of sight and hence no direct ray)









## Single carrier modulation

Lets suppose a single carrier BPSK modulation

- Carrier frequency,  $f_c = 1/T_c$ ; 1 bit per symbol
  - $b_i = 1$ , 180° degrees phase shift
  - b<sub>i</sub> = 0, o<sup>o</sup> degrees phase shift

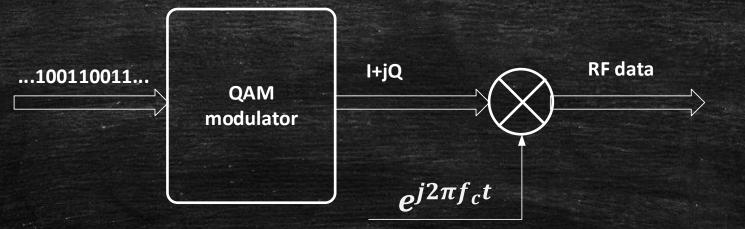
Tc

Ts

• Bit rate  $R = 1/T_s$ , where  $T_s$  represents the symbol time

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A simple transmitter example



0	0	0	0	0	0	0	0	0	0	•	0	0	0	
						0	0	0	0	0	0	0	•	
	Q	0	0	0	°q	0	0	•	•	0	0	•	0	
	<u> </u>				<u> </u>	•	0	•	0	0	0	•	<u>° Q</u>	
		0	0	0	0	0	0	0	0	0	0	•	0	
						0	0	0	0	0	0	0	0	
0	0	0	0	0	0	0	•	0	0	0	0	0	0	
						0	0	•	0	0	0	0	0	
ODSK		16 0 0 0				64-QAM								
QPSK		16-QAM												

### Different echoes effect:

Transmitted data

 $S_1$ 

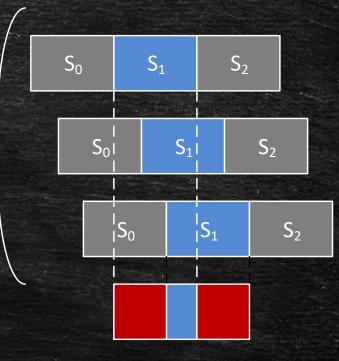
S<sub>0</sub>

Propagation through the channel

 $\tau_0 \tau_1$ 

てっ

At the receiver



- The red coloured part is affected by ISI

 $S_2$ 

- Interference produced by other symbols
- The symbol at the receiver seems longer in time
- In the following part we will focus on the blue part
  - Only interference from the own symbol delayed

ρ

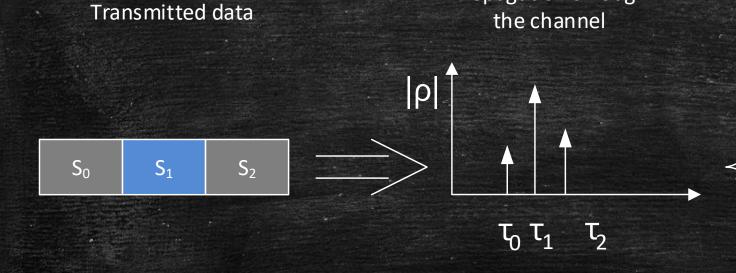
 All the stated is observed in the time domain, but what is its interpretation in the frequency domain?

 $S(t) \stackrel{FFT}{\longleftrightarrow} S(f)$  $S(t-\tau) \stackrel{FFT}{\longleftrightarrow} e^{-j2\pi f\tau} S(f)$ 

• For a channel as the one shown in the previous example:  $S_{Rx}(f) = \rho_0 S(f) e^{-j2\pi f\tau_0} + \rho_1 S(f) e^{-j2\pi f\tau_1} + \rho_2 S(f) e^{-j2\pi f\tau_2}$   $S_{Rx}(f) = S(f) \left(\rho_0 e^{-j2\pi f\tau_0} + \rho_1 e^{-j2\pi f\tau_1} + \rho_2 e^{-j2\pi f\tau_2}\right)$   $S_{Rx}(f) = S(f) H(f)$ 

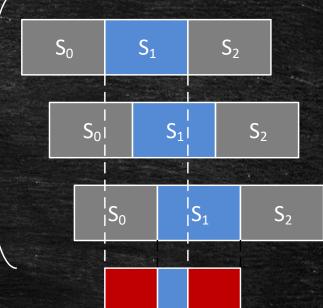
H(f) represents the channel response in the frequency domain

- The channel effect can produce high performance loss
  - The combination of the different paths can be constructive or destructive
  - The destructive combinations can lead to huge attenuation (fadings) even to erasure events
  - The channel equalization is necessary to solve the problems associated to the channel fadings and erasures



#### Now we will focus on the red part of the received symbol Propagation through

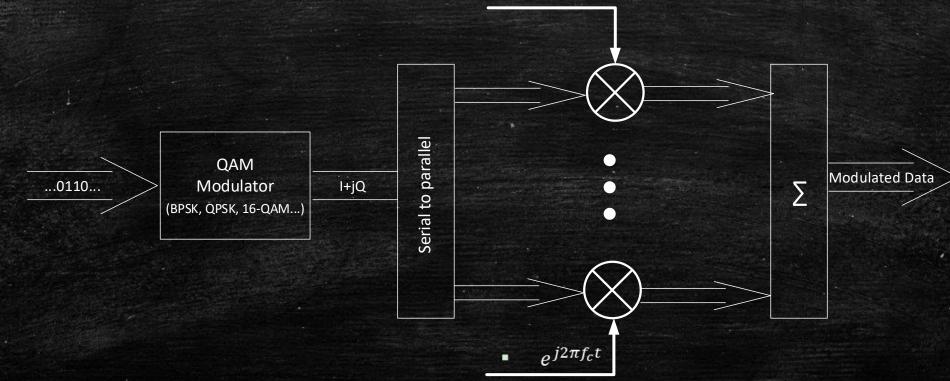
At the receiver



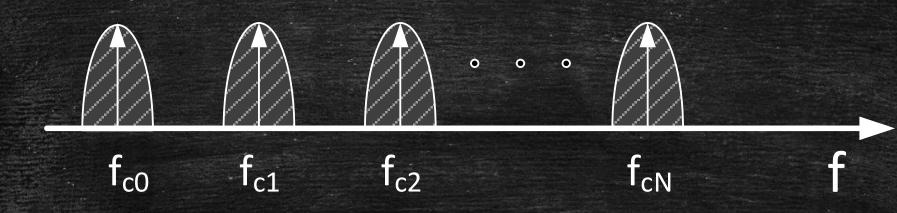
- Sum of different delayed symbols
- In order to recover the information the symbol
  - T<sub>s</sub> must be much longer than the difference between the first ray and the last one of the channel (\(\tau\_{max})\)
  - The longer the symbol is the less rate we obtain

## Multi carrier modulation

- The solution for the aforementioned problem is parallelize
  - Send the information through different carriers with higher symbol period



The spectrum at the output of the modulator:



- For a N+1 carriers, input data rate R, and BPSK modulation
  - For a same data rate R, the symbol time of a multi-carrier is N+1 times higher than for a single carrier
  - For a given  $au_{max}$  the data rate is higher than in a single carrier
  - The spectrum occupied is much wider!

## Orthogonal Frequency Division Multiplexion

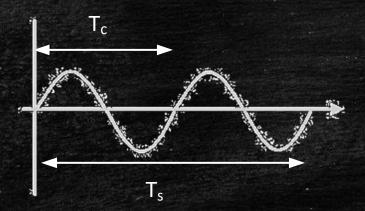
Spectral characteristics of a FDM signal:

•  $\phi_c = e^{j2\pi f_c t} rect\left(\frac{t}{T_s}\right) = \begin{cases} e^{j2\pi f_c t} & \text{if } 0 \le t \le T_s \\ 0 & \text{in any other case} \end{cases}$ 

Time (real part):

Frequency:

 $1/T_s$ 



- If the N carriers in a OFDM symbol satisfy:

   f<sub>ck</sub> = k/T<sub>c</sub> with k = -N/2, ..., 0, ..., N/2 1
- The carriers are orthogonal
  - The maximum of a carrier coincides with the zeros of the others

 $1/T_c$   $2/T_c$ 

 $-2/T_{c} - 1/T_{c} = 0$ 

Mathematically the expression for a determinate symbol:

$$s(t) = \sum_{k=0}^{\infty} a_k e^{j2\pi f_{ck}t} rect\left(\frac{t}{T_s}\right)$$

-  $a_k$  represents the k-th output symbol of the QAM modulator ( $a_k = I_k + jQ_k$ ) - N represents the number of carriers in the OFDM symbol -  $f_{ck} = k/T_c$  with k = 0, 1, ..., N - 1

The problem is now obtaining that many analogue oscillators!

- In the digital domain, if sampled with N samples, a symbol:  $T_{samp} = T_s/N$
- The time becomes discrete,  $t = nT_{samp}$ , and then:

$$f_{ck}t = \left(\frac{k}{T_S}\right)n(T_S/N) = kn/N$$
$$s[n] = \sum_{\substack{N=1\\N \notin \overline{1}^0}}^{N-1} a_k e^{\frac{j2\pi kn}{N}} rect \left[\frac{n-\frac{N}{2}}{N}\right]$$
$$s[n] = \sum_{\substack{k=0\\k=0}}^{N \notin \overline{1}^0} a_k e^{j\frac{2\pi}{N}kn} n \in [0, N-1]$$

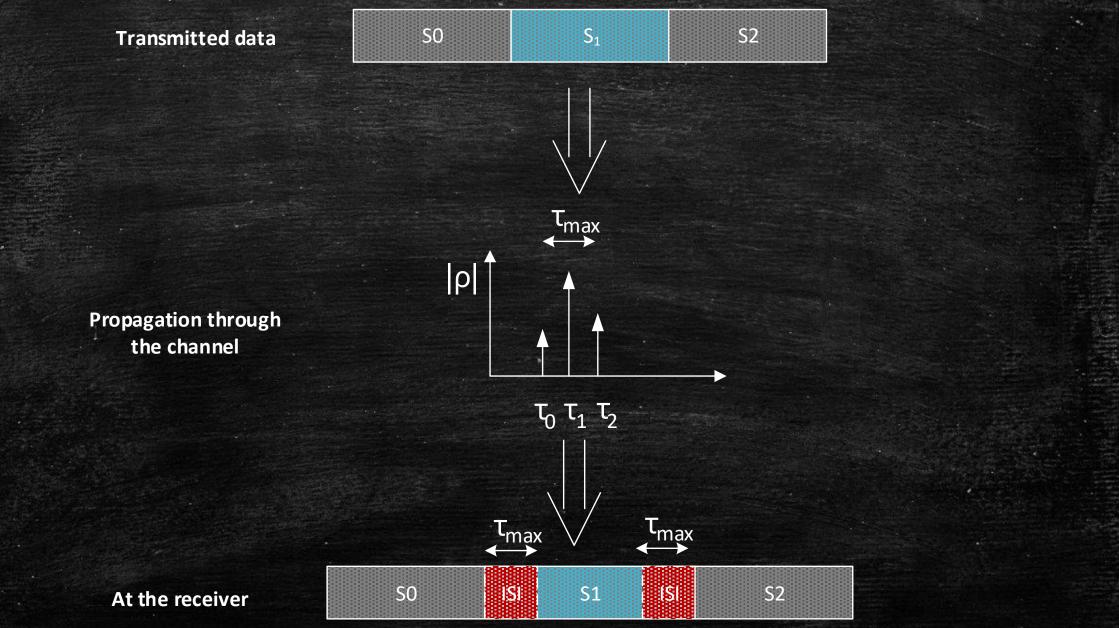
Looking carefully to the expression achieved before:

- It matches with the IDFT of  $a_k!$ 

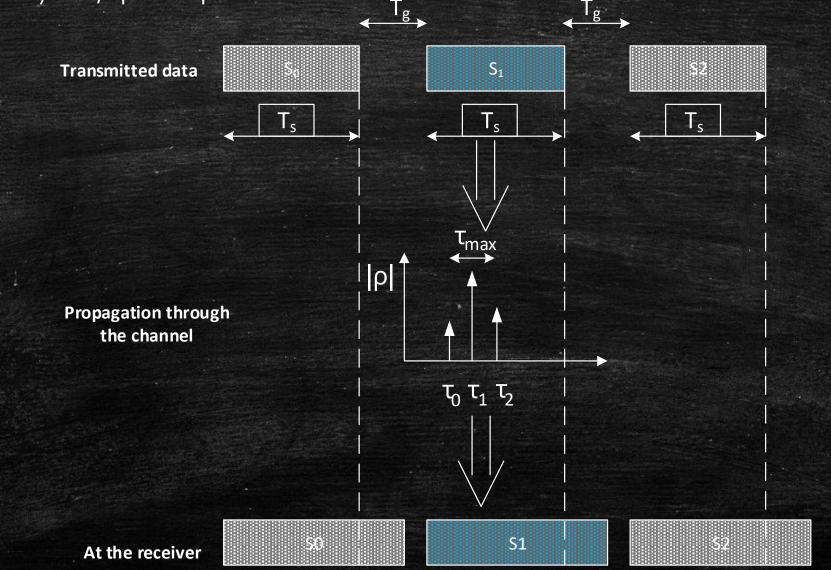
$$s[n] = \sum_{k=0}^{N-1} a_k e^{j\frac{2\pi}{N}kn} = IDFT[a_k]$$

- Every  $a_k$  represents the amplitude and phase of the carriers
- We are coming form the frequency domain to the time domain
  - Every block previous to the IDFT in an OFDM system belongs to the frequency domain
  - Every block after the IDFT in an OFDM system belongs to the time domain

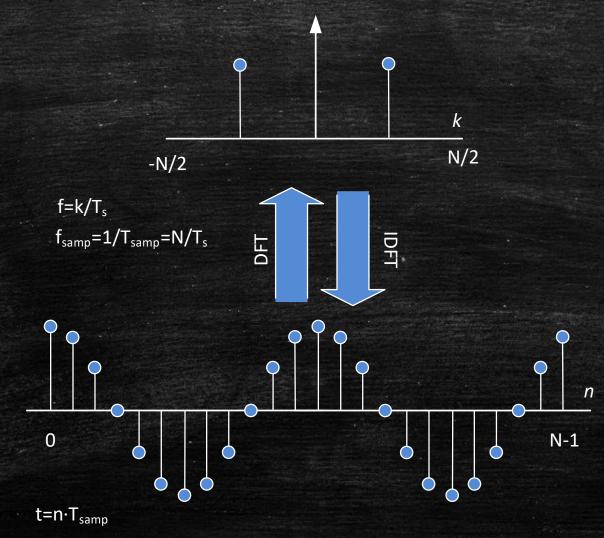
#### There is still a problem to be solved: ISI



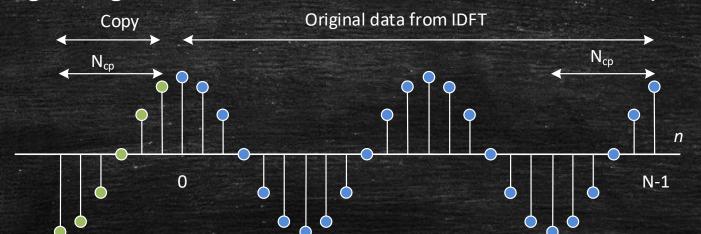
- With a longer symbol time the ISI is reduced (in %) but the problem still exists
  - A possible solution is to insert a guard period between symbols
    - Capacity loss, spread spectrum



- Another way to cope with ISI is the cyclic prefix
  - DFT and IDFT are applied to periodic signals
  - The result of the IDFT is a period of a periodic signal in time

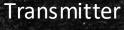


 The cyclic prefix consists in taking the last N<sub>cp</sub> samples of the signal and copy it at the beginning of the symbol at the receiver these samples are discarded



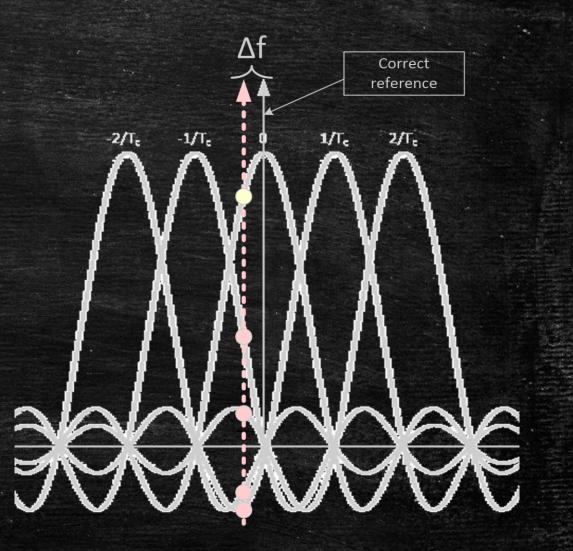
Reduces the capacity but doesn't spread the spectrum







- There is an additional problem with OFDM
  - The frequency reference in both transmitter and receiver are not usually exactly same (oscillators deviations, mismatching of the elements ...)
  - This leads to a orthogonality loss because the sampling instant is not exactly in the maximum (and cero of the other carriers)
  - This is additional interference coming from other carriers (ICI)



# THANKS!

Any questions?

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