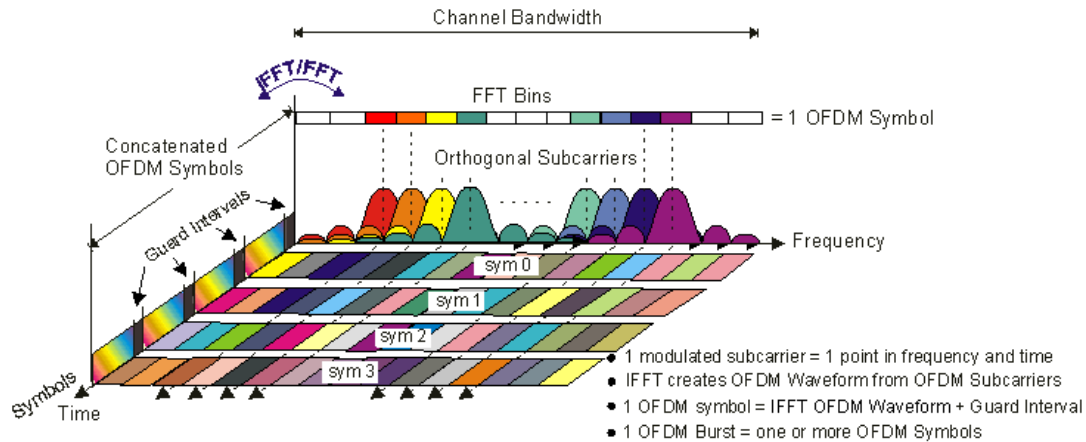


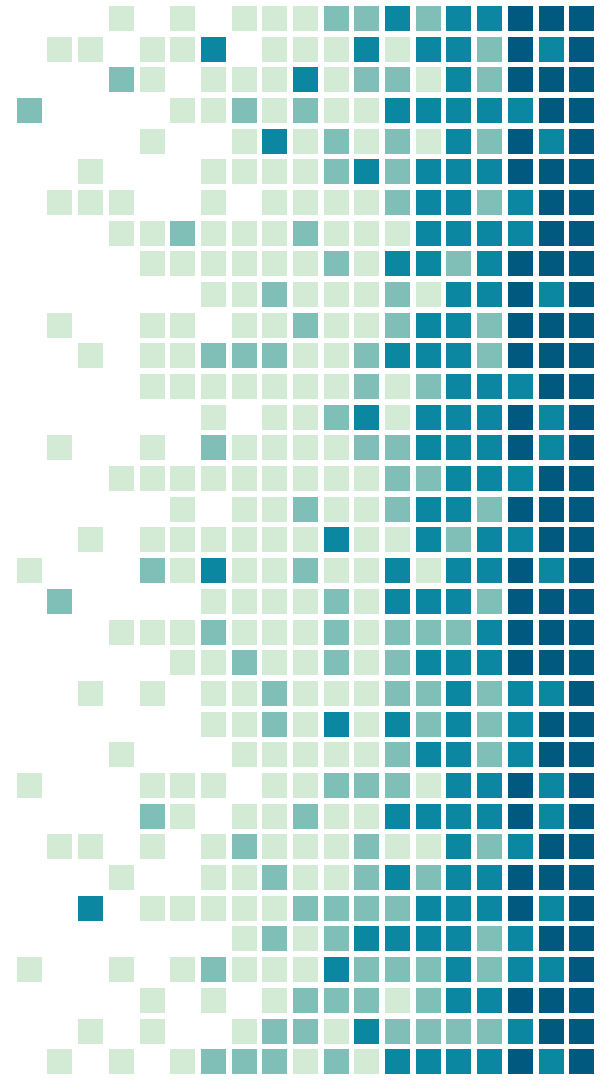
# OFDM

## Basic concepts

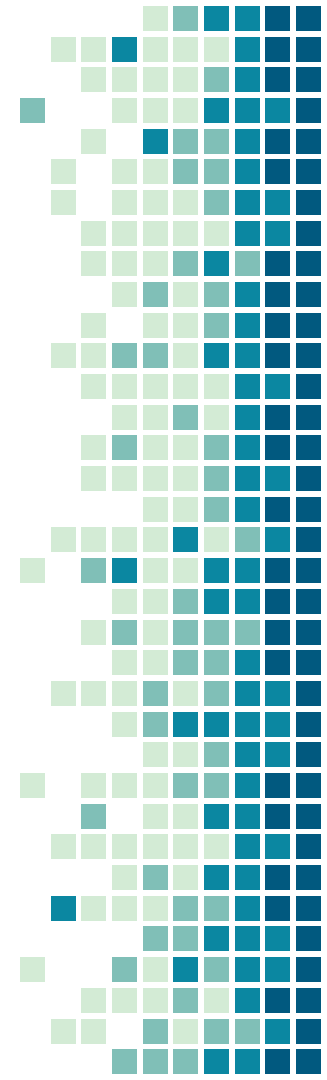


# 1. Introduction

General knowledge

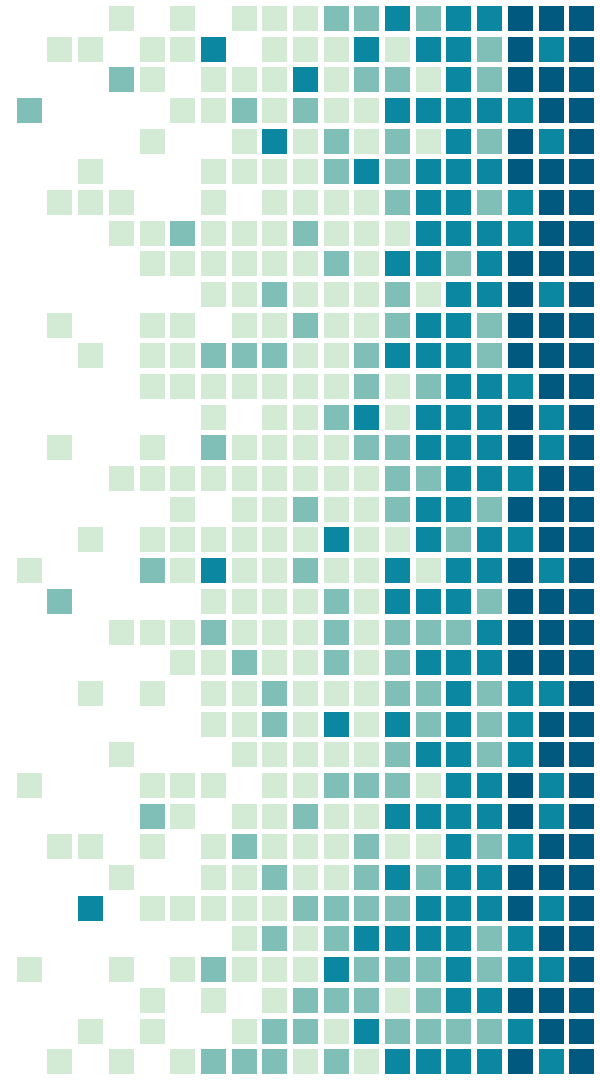


- Orthogonal Frequency Division Multiplexion
  - A number of orthogonal carriers bear the information
  - Resilient to time synchronization errors
  - Very high spectral efficiency
  
- OFDM is a wide used transmission technique
  - ADSL
  - IEEE 802.11a/g/n and WiMAX
  - DVB-T , DVB-H , DVB-T2, DVB-NGH



# 2. Propagation scenarions

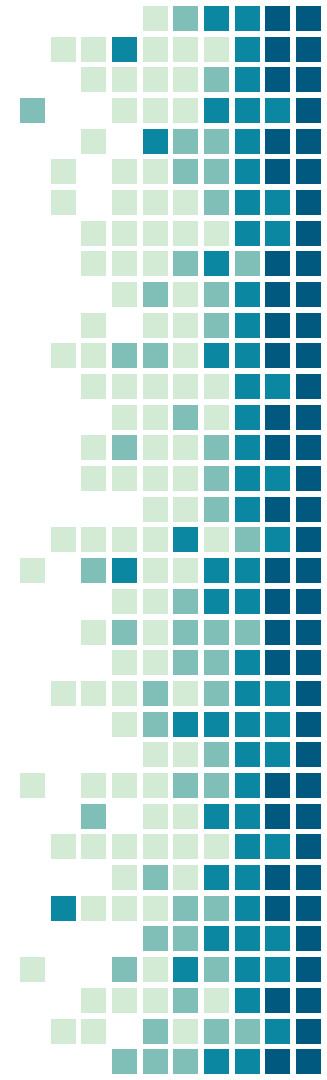
Why do we use OFDM?



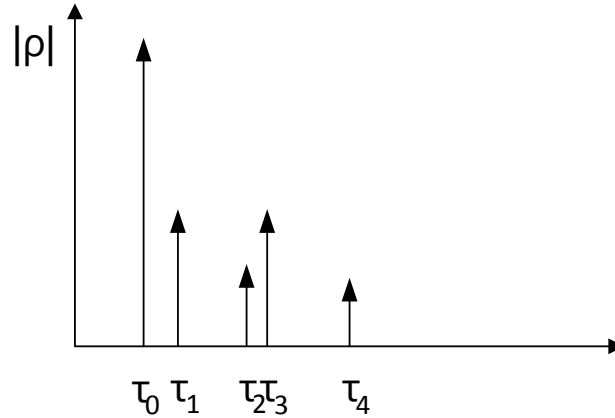
- In wireless communications there are much harder propagation conditions than in wired communications
- There are different propagations paths from the transmitter to the receiver (multi path propagation)



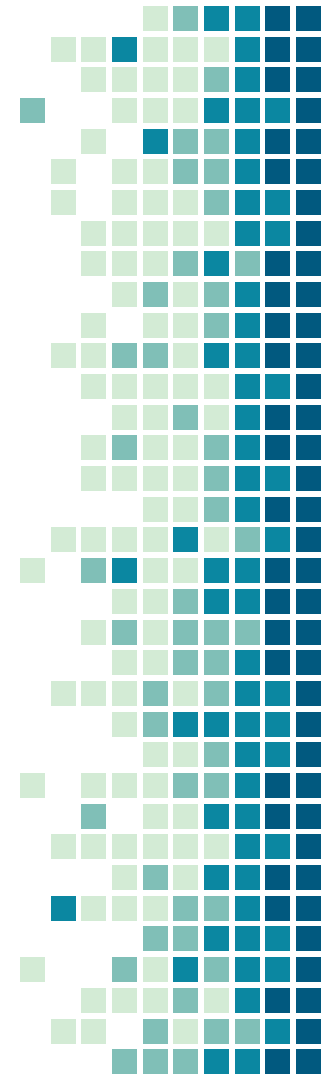
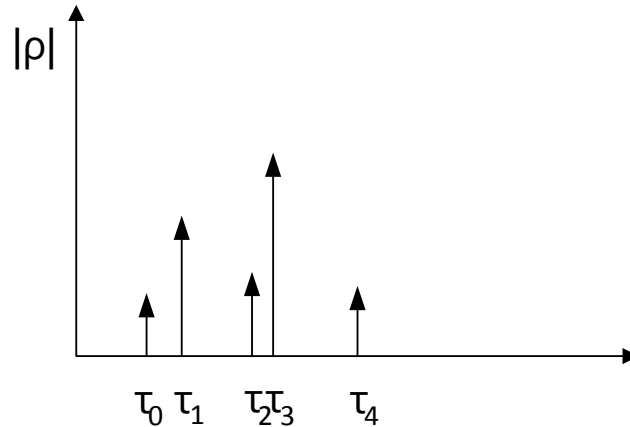
- The propagation scenarios are usually modelled as a FIR filter which taps represent:
  - Delay ( $\tau$ )
  - Path complex gain ( $\rho$ )
- The propagation scenarios can be divided in two main types
  - Ricean (there is line of sight and hence direct ray)
  - Rayleigh (there is no line of sight and hence no direct ray)



- The Ricean channel looks like:

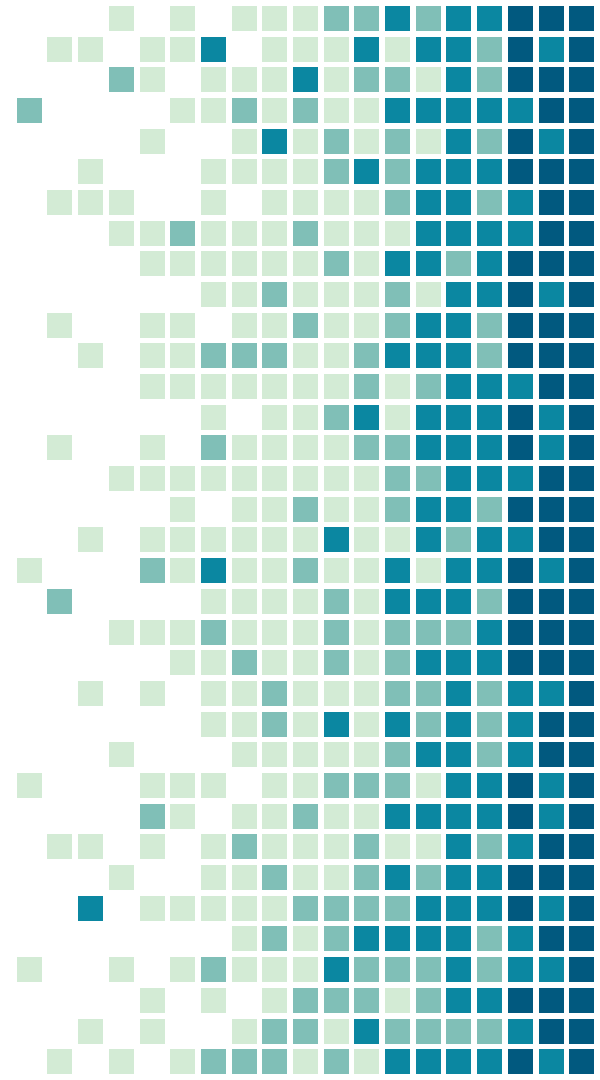


- The Rayleigh channel looks like:



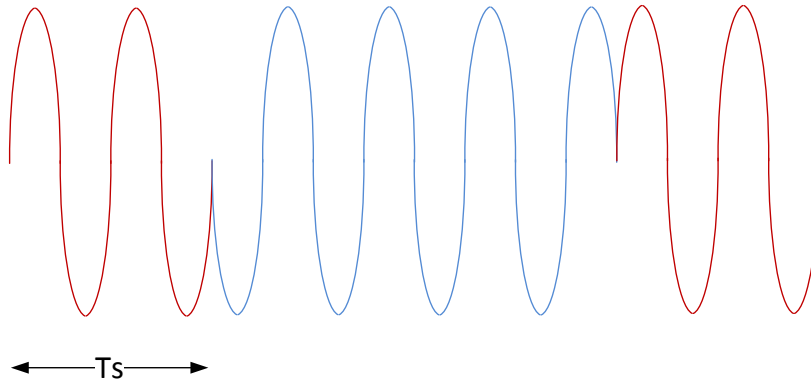
# 3. Single carrier modulation

Traditional approach for  
communication systems

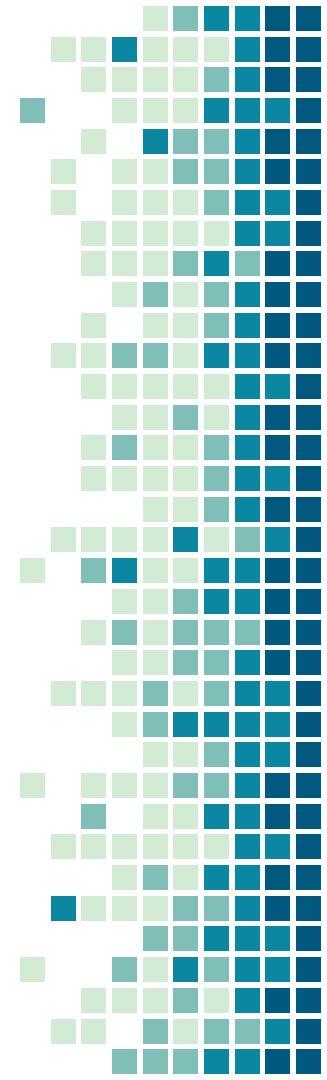
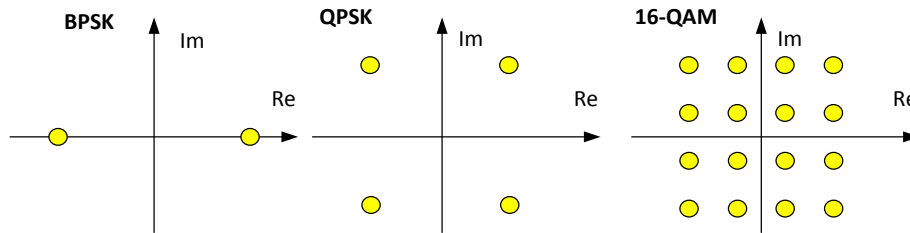
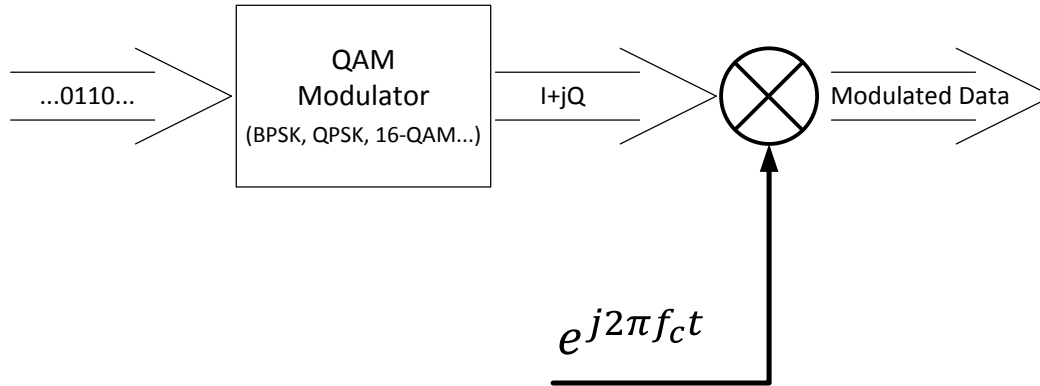




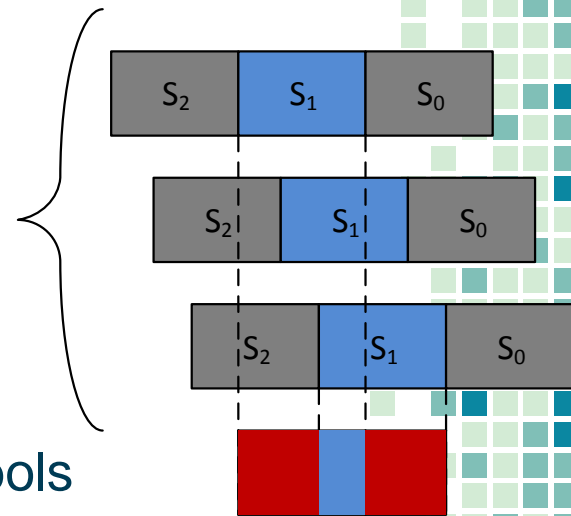
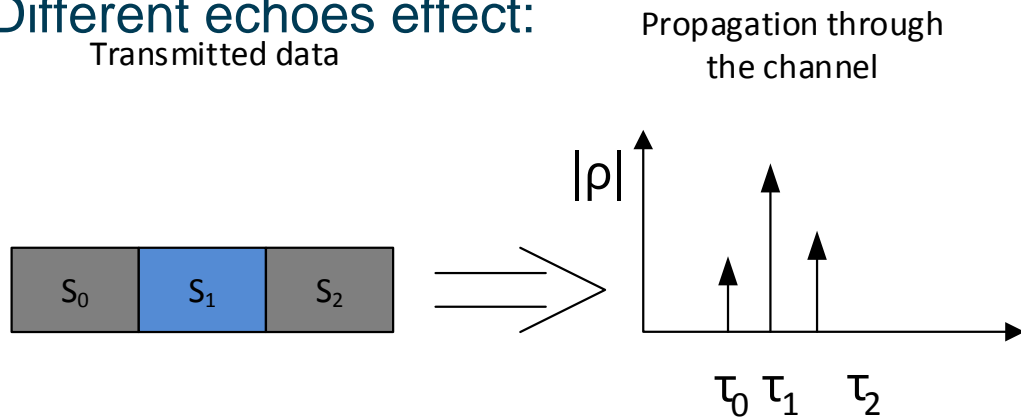
- Lets suppose a single carrier BPSK modulation
  - Carrier frequency,  $f_c = 1/T_c$
  - 1 bit per symbol
    - $b_i = 1$ ,  $180^\circ$  degrees phase shift
    - $b_i = 0$ ,  $0^\circ$  degrees phase shift
    - Bit rate  $R = 1/T_s$ , where  $T_s$  represents the symbol time



- A basic transmission scheme:



■ Different echoes effect:  
Transmitted data



- The red coloured part is affected by ISI
  - Interference produced by other symbols
- The symbol at the receiver seems longer in time
- In the following part we will focus on the blue part
  - Only interference from the own symbol delayed

- All the stated is observed in the time domain, but what is its interpretation in the frequency domain?

$$S(t) \xleftrightarrow{FFT} S(f)$$

$$S(t - \tau) \xleftrightarrow{FFT} e^{-j2\pi f\tau} S(f)$$

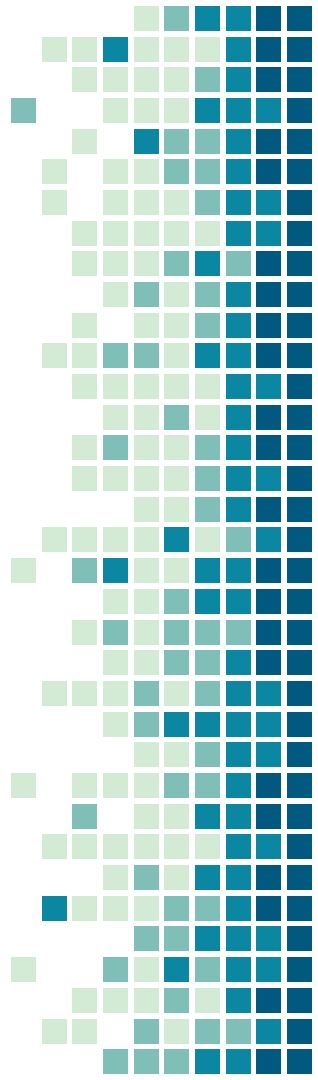
- For a channel as the one shown in the previous example:

$$S_{Rx}(f) = \rho_0 S(f) e^{-j2\pi f\tau_0} + \rho_1 S(f) e^{-j2\pi f\tau_1} + \rho_2 S(f) e^{-j2\pi f\tau_2}$$

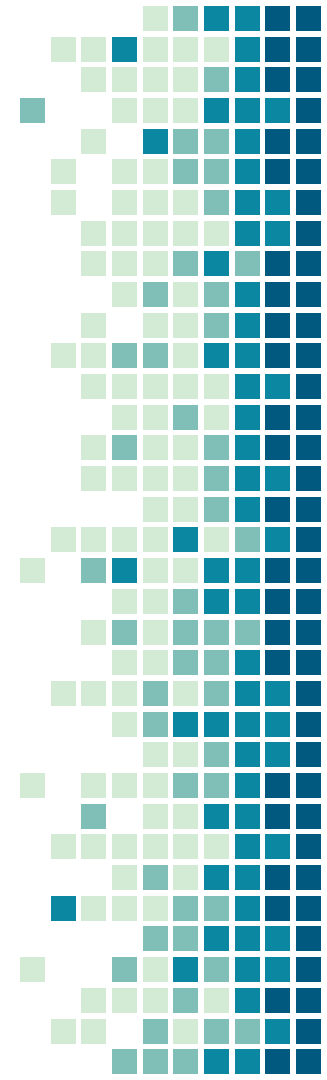
$$S_{Rx}(f) = S(f) (\rho_0 e^{-j2\pi f\tau_0} + \rho_1 e^{-j2\pi f\tau_1} + \rho_2 e^{-j2\pi f\tau_2})$$

$$S_{Rx}(f) = S(f) H(f)$$

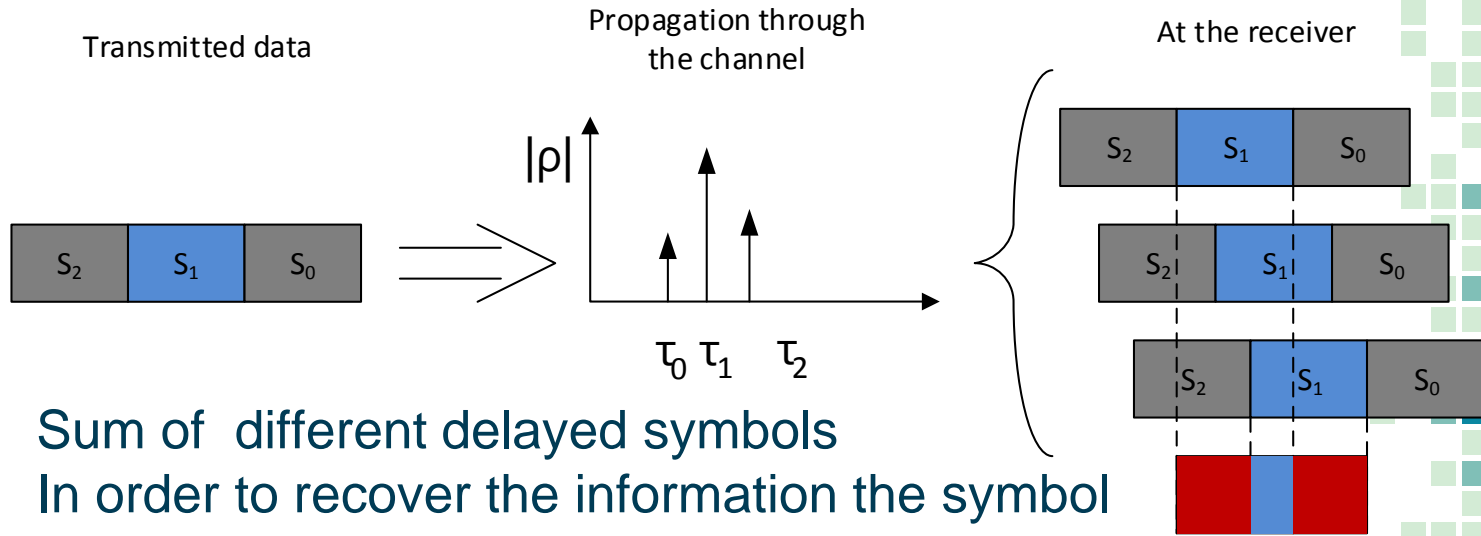
- $H(f)$  represents the channel response in the frequency domain



- The channel effect can produce high performance loss
  - The combination of the different paths can be constructive or destructive
  - The destructive combinations can lead to huge attenuation (fadings) even to erasure events
  - The channel equalization is necessary to solve the problems associated to the channel fadings and erasures



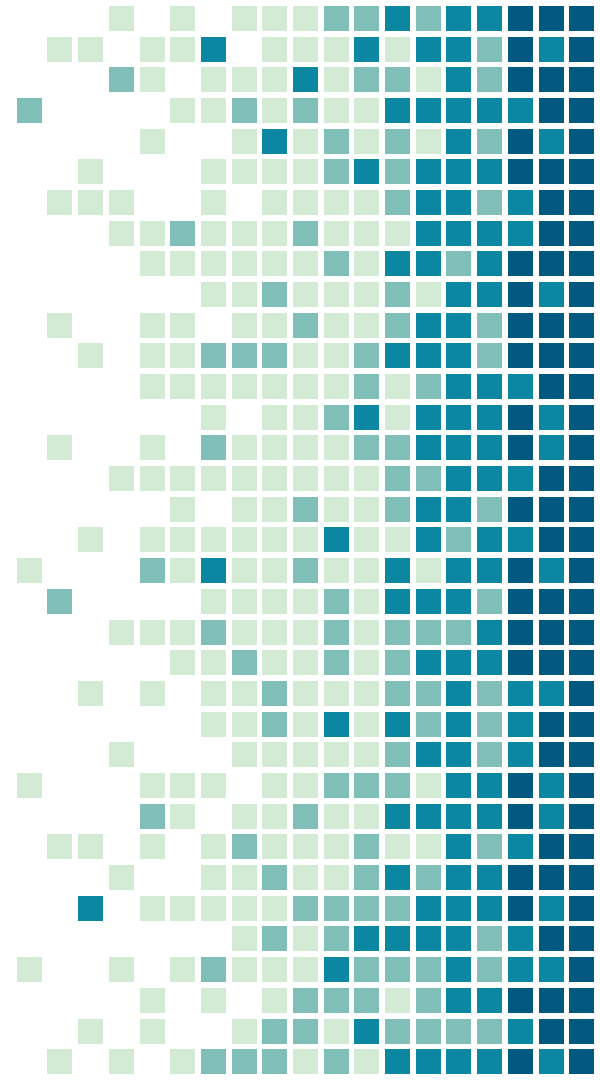
- Now we will focus on the red part of the received symbol



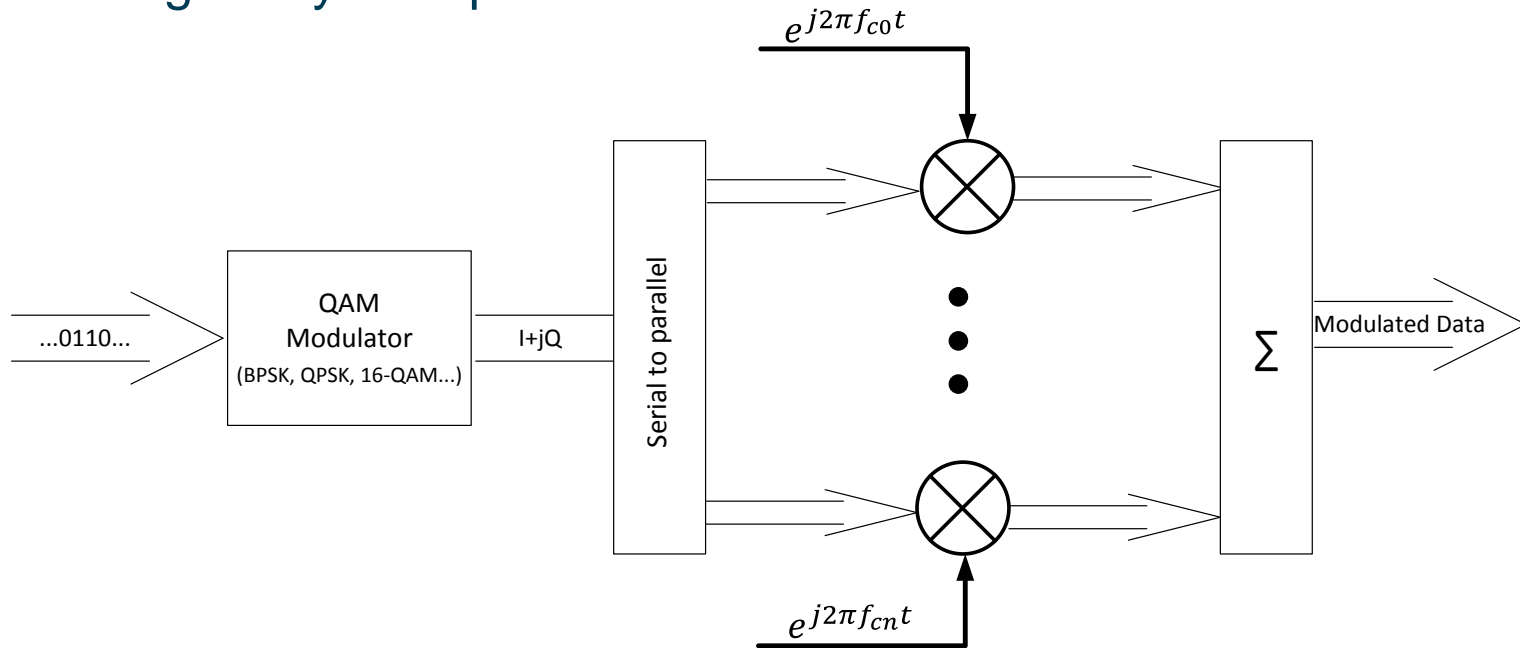
- Sum of different delayed symbols
- In order to recover the information the symbol
  - $T_s$  must be much longer than the difference between the first ray and the last one of the channel ( $\tau_{max}$ )
  - The longer the symbol is the less rate we obtain

# 4. Multi carrier modulation

How to ease ISI

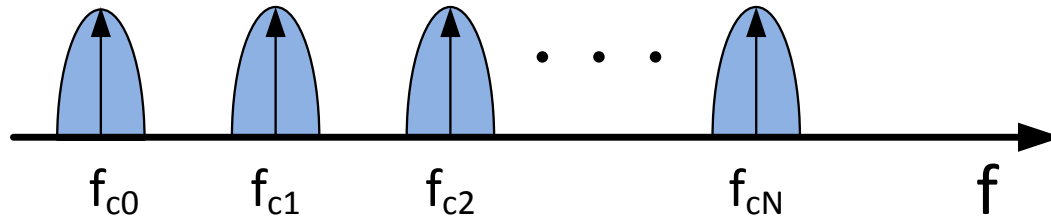


- The solution for the aforementioned problem is parallelize
  - Send the information through different carriers with higher symbol period





- The spectrum at the output of the modulator:

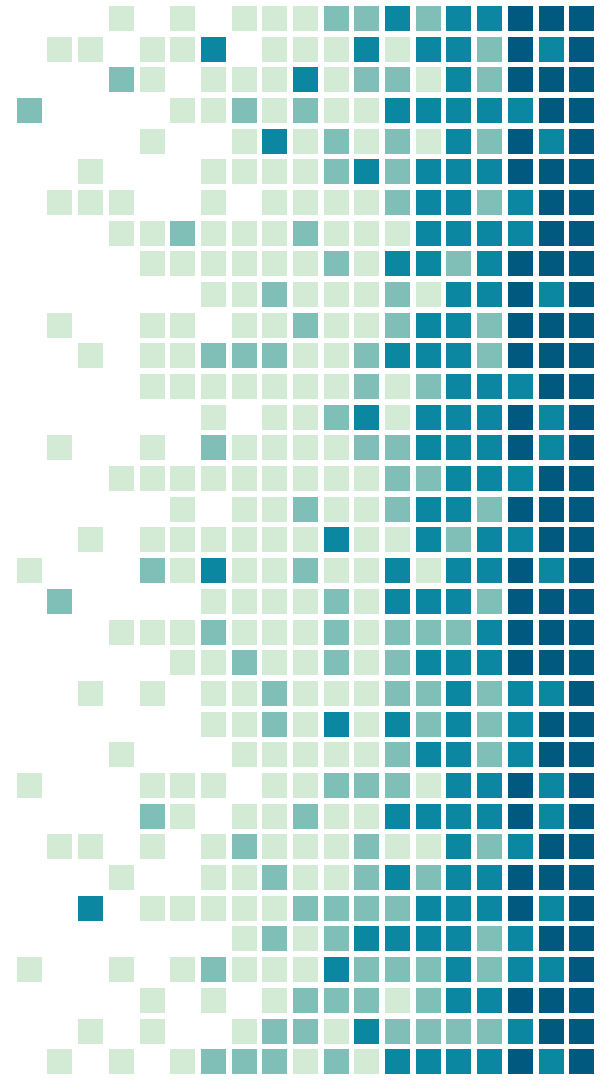


- For a  $N+1$  carriers, input data rate  $R$ , and BPSK modulation
  - For a same data rate  $R$ , the symbol time of a multi-carrier is  $N+1$  times higher than for a single carrier
  - For a given  $\tau_{max}$  the data rate is higher than in a single carrier
  - The spectrum occupied is much wider!



# 5. Orthogonal Frequency Division Multiplexion

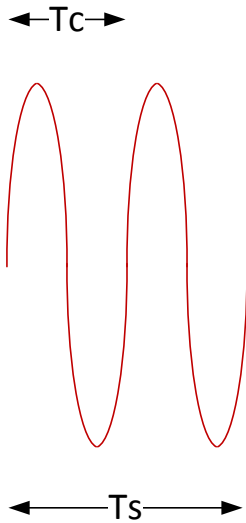
OFDM basics



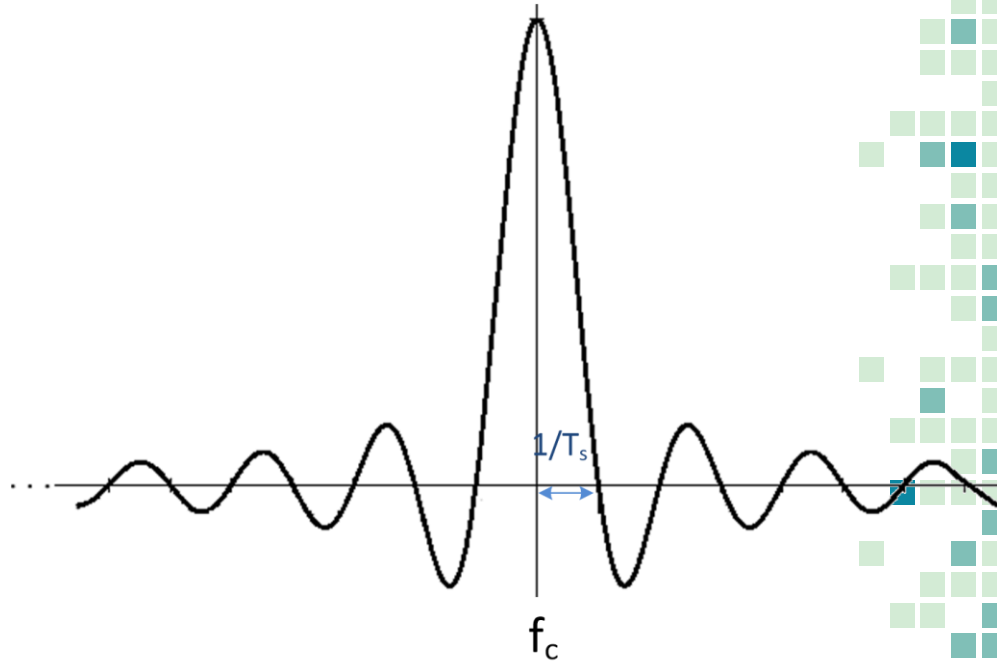
- Spectral characteristics of a FDM signal:

- $$\phi_c = e^{j2\pi f_c t} \text{rect}\left(\frac{t - \frac{T_s}{2}}{T_s}\right) = \begin{cases} e^{j2\pi f_c t} & \text{if } 0 \leq t \leq T_s \\ 0 & \text{in any other case} \end{cases}$$

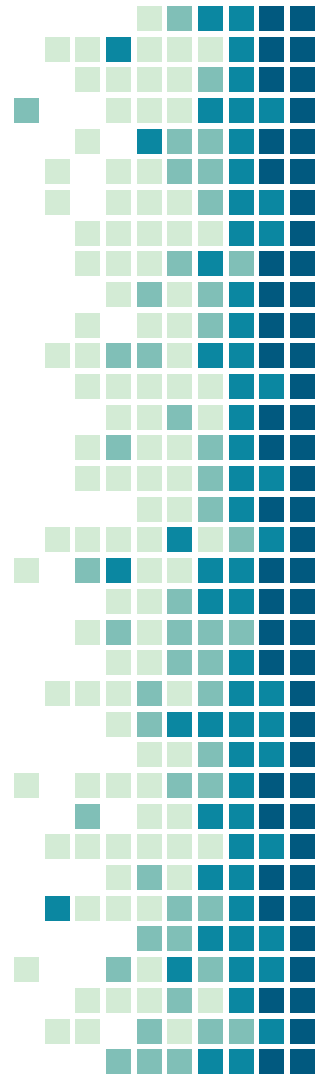
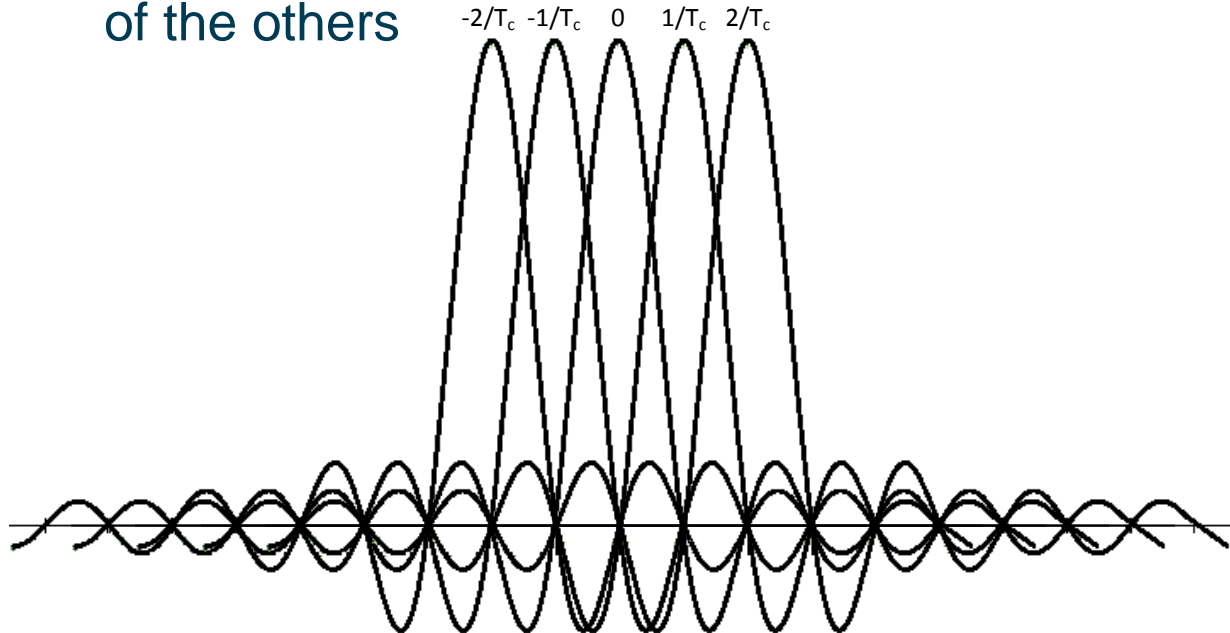
- Time (real part):



Frequency:



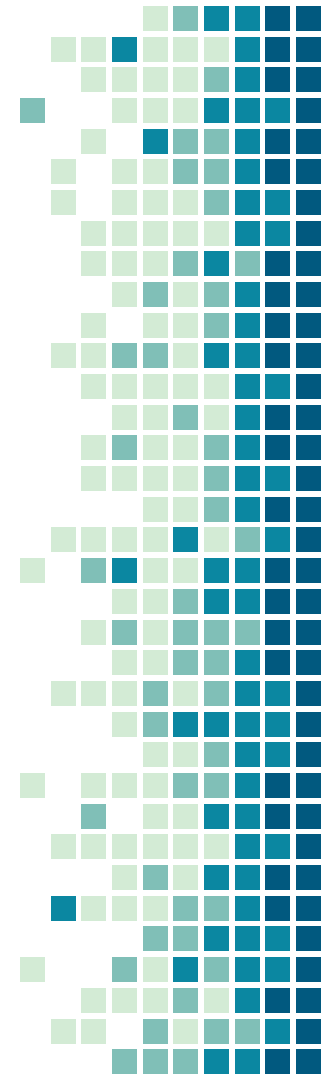
- If the  $N$  carriers in a OFDM symbol satisfy:
  - $f_{ck} = k/T_c$  with  $k = -N/2, \dots, 0, \dots, N/2 - 1$
- The carriers are orthogonal
  - The maximum of a carrier coincides with the zeros of the others



- Mathematically the expression for a determinate symbol:

$$s(t) = \sum_{k=0}^{N-1} a_k e^{j2\pi f_{ck}t} \text{rect}\left(\frac{t - \frac{T_s}{2}}{T_s}\right)$$

- $a_k$  represents the k-th output symbol of the QAM modulator ( $a_k = I_k + jQ_k$ )
  - N represents the number of carriers in the OFDM symbol
  - $f_{ck} = k/T_c$  with  $k = 0, 1, \dots, N - 1$
- The problem is now obtaining that many analogue oscillators!



- In the digital domain, if sampled with  $N$  samples, a symbol:

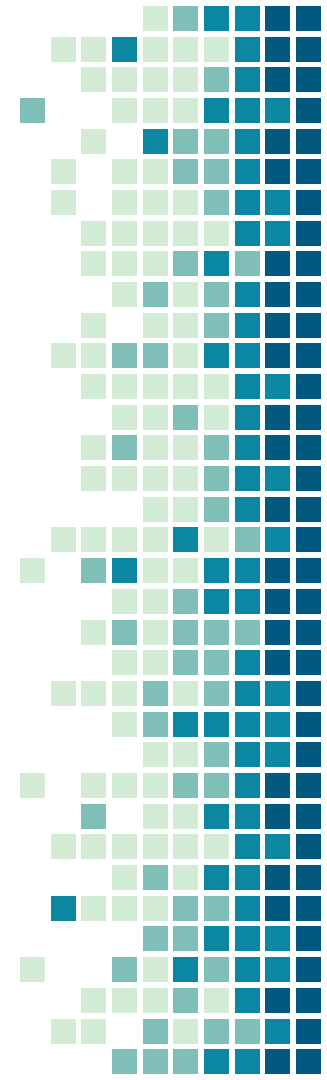
$$T_{s\text{amp}} = T_s/N$$

- The time becomes discrete,  $t = nT_{s\text{amp}}$ , and then:

$$f_{ck}t = \left(\frac{k}{T_s}\right)n(T_s/N) = kn/N$$

$$s[n] = \sum_{k=0}^{N-1} a_k e^{\frac{j2\pi kn}{N}} \text{rect}\left[\frac{n - \frac{N}{2}}{N}\right]$$

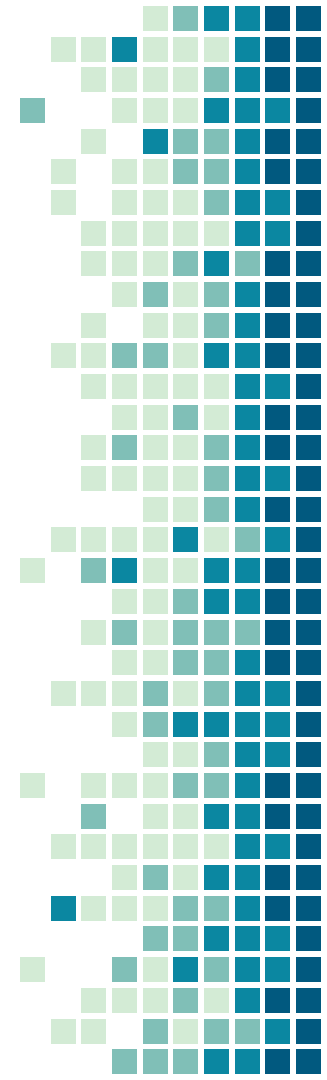
$$s[n] = \sum_{k=0}^{N-1} a_k e^{j\frac{2\pi}{N}kn} \quad n \in [0, N - 1]$$



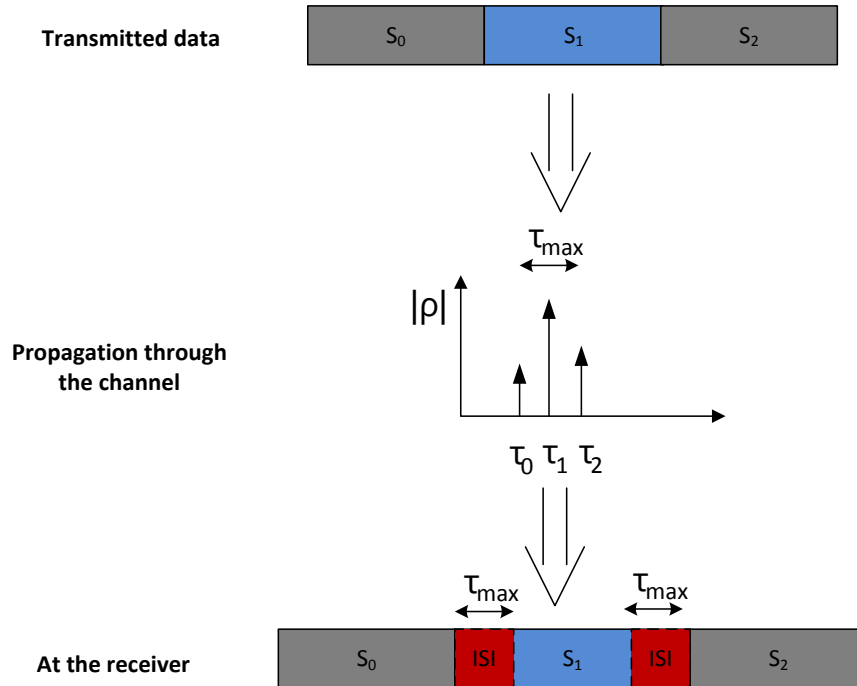
- Looking carefully to the expression achieved before:
  - It matches with the IDFT of  $a_k$ !

$$s[n] = \sum_{k=0}^{N-1} a_k e^{j\frac{2\pi}{N}kn} = IDFT[a_k]$$

- Every  $a_k$  represents the amplitude and phase of the carriers
- We are coming from the frequency domain to the time domain
  - Every block previous to the IDFT in an OFDM system belongs to the frequency domain
  - Every block after the IDFT in an OFDM system belongs to the time domain

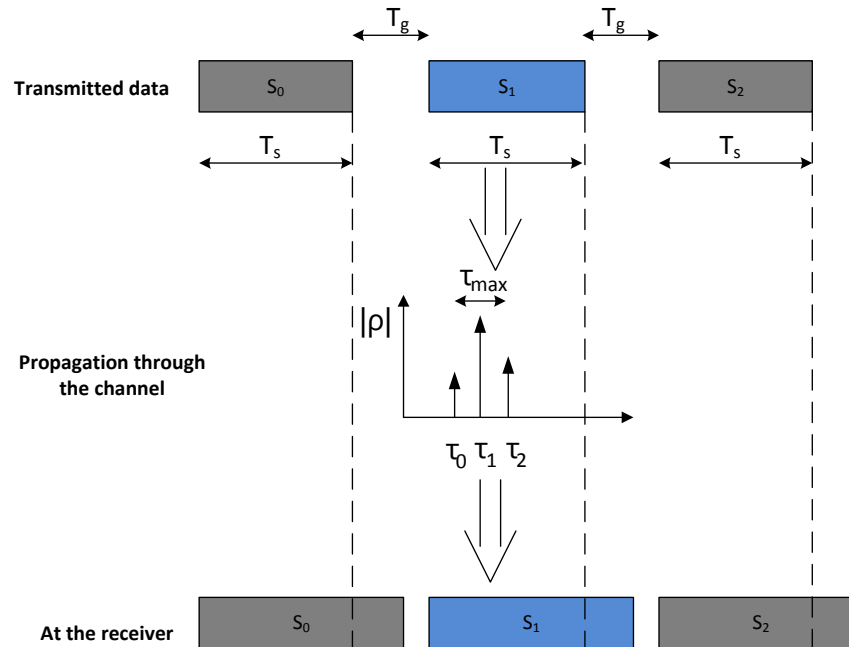


- There is still a problem to be solved: ISI

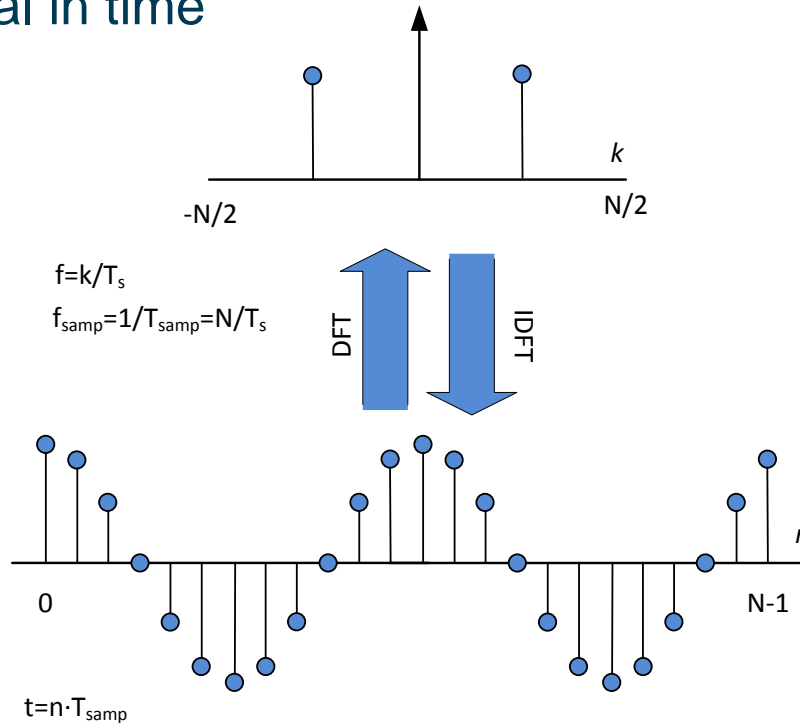




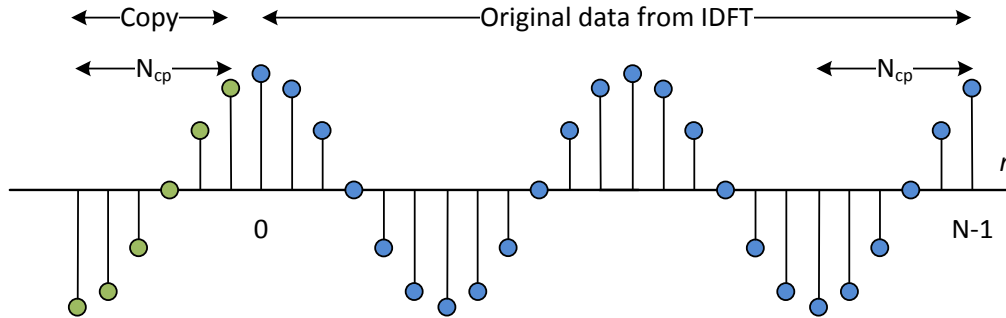
- With a longer symbol time the ISI is reduced (in %) but the problem still exists
  - A possible solution is to insert a guard period between symbols
    - Capacity loss, spread spectrum



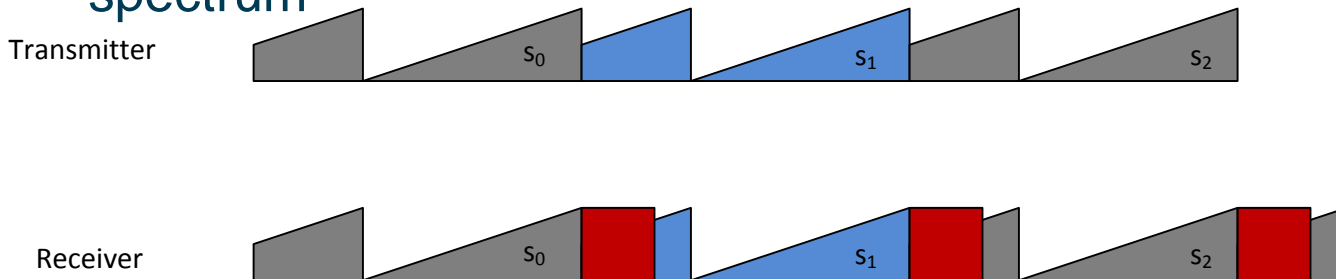
- Another way to cope with ISI is the cyclic prefix
  - DFT and IDFT are applied to periodic signals
  - The result of the IDFT is a period of a periodic signal in time



- The cyclic prefix consists in taking the last  $N_{cp}$  samples of the signal and copy it at the beginning of the symbol at the receiver these samples are discarded

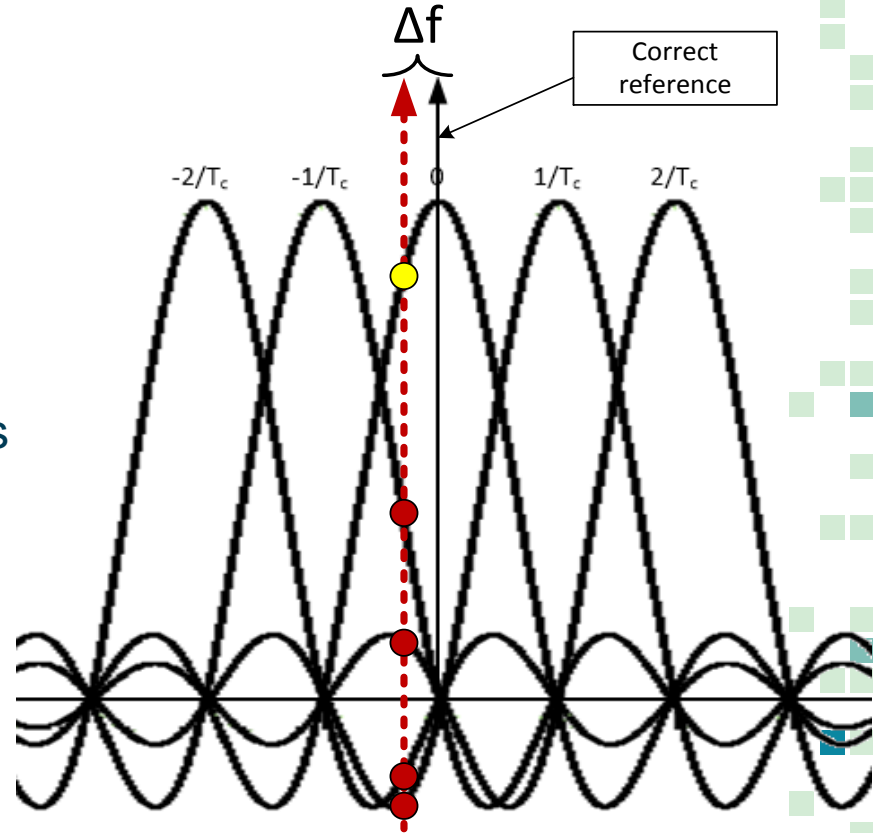


- Reduces the capacity but doesn't spread the spectrum



- There is an additional problem with OFDM

- The frequency reference in both transmitter and receiver are not usually exactly same (oscillators deviations, mismatching of the elements ...)
- This leads to a orthogonality loss because the sampling instant is not exactly in the maximum (and zero of the other carriers)
- This is additional interference coming from other carriers (ICI)



# THANKS!

Any questions?

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